

Motor-In-The-Loop Approach in the Validation of an Induction Machine Model – Online Rotor Resistance Estimation

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Abstract: In The article introduces a new motor-in-the-loop (MotIL) concept as a new system level test method for variable frequency drives. The main idea of this methodology is to use a slip ring motor to investigate the effect of the drive system on a real motor. Because of the slip ring structure, the rotor side signals and parameters can be directly observed online as well as offline. Therefore, it can be a good extension of the HIL and PHIL solutions. Compared with HIL and PHIL methodology the MotIL makes it possible to observe the drive system behavior in real physical environment including noises, coupling, nonlinearity and other potentially not modeled factors.

A case study related to rotor resistance estimation is also presented aiming to prove the performance of the MotIL concept. The presented MotIL algorithm focuses on the accurate and robust rotor flux calculation and online rotor resistance measurement. The concept of the introduced algorithm is based on collaboration and self-adaptation of two independent rotor flux estimator.

Keywords: Rotor Resistance, Estimation, HIL, FOC

INTRODUCTION

The usage of induction machines in applications where high torque accuracy and dynamic behavior are critical typically requires the field oriented control (FOC) [1], [2].

The hardware-in-the-loop (HIL) and power HIL (PHIL) solutions [3], [4], [5] make it possible to test the FOC based motor controller in its real embedded and power circuit thanks to the fact that they can emulate effect of the load. The HIL makes possible to test the electronic control unit (ECU) by means of the real time model of the power parts and sensor system. The test of integrated power system's behavior is possible with the PHIL. The efficacy of these verification concept strongly depends on the real time plant model fidelity in the HIL or PHIL.

To handle the risk related to model fidelity, a physical motor based test framework with high level of observability, accuracy and configurability would be a promising solution. Based on the slip ring induction machine (SRIM) this framework can be easily created, this arrangement would be referred to in this article as motor-in-the-loop (MotIL). The main advantage of MotIL is that the motor behavior is not simulated. An additional benefit is a high-level of observability of the rotor side signals. Therefore, direct validation of the motor model in the FOC controller is also possible. Since MotIL framework cannot be so generic as the PHIL, therefore it is not an alternative solution rather an extension.



Fig. 1. Signal flow block diagram of the developed algorithm.

The current article introduces a novel approach to validate and monitor FOC based drives. It mainly covers the verification of the accuracy of the motor model since it strongly influences the performance of the motor controller. In order to demonstrate the efficacy of the methodology, online rotor resistance estimation was chosen as a case study.

There are many online and offline solutions for estimating the rotor resistance because of its importance in the FOC. [6] and [7] use the back-EMF for the parameter identification. [8] shows an online solution to estimate the rotor resistance. [9] uses the reactive power for the estimation of the rotor

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resistance. Kalman Filter also can be used to estimate the rotor resistance [10], [11]. [12] shows a method for estimating both the rotor and stator resistances. Compared with recommended MotIL concept, most of these techniques require signal injection by the inverter.

THE DEVELOPED ALGORITHM

The presented Model Reference Adaptive System based online rotor resistance measurement process uses on two rotor flux $\overline{\Psi}'_{rSRef}^{xy}$ (rotor flux space vector based on rotor current in stator reference frame) and $\overline{\Psi}'_{rS}^{xy}$ (rotor flux space vector transformed to stator side in stator reference frame) estimation methods. The first one uses the rotor side measurements so it has no need to use the rotor resistance to calculate rotor flux (Fig. 1). The second one uses the traditional motor model (MM) (Fig. 1) which estimates the rotor flux based on the rotor mechanical speed and stator currents. Since, only the second one uses the input for the rotor resistance tuning.

A. Rotor current based rotor flux estimation

The rotor current measurement in case of a SRIM is possible and this allows the calculation of the rotor flux without the knowledge the rotor resistance. As long as the effect of the saturation can be neglected this calculation will result in a correct rotor flux that can be accepted as a reference. This flux is used to adjust the rotor flux calculated by the machine model.

Equation (1) defines the flux equation of the induction machine in stationary reference frame:

$$\bar{\Psi}'_{rS}{}^{xy} = \bar{\Psi}_{sS}{}^{xy} - \bar{\Psi}_{slS}{}^{xy} + \bar{\Psi}'_{rlS}{}^{xy} \tag{1}$$

where

$$\overline{\Psi}_{ss}^{xy} = \int \left(\overline{U}_{ss} - \overline{\iota}_{ss}^{xy} R_s \right) dt \tag{2}$$

$$\overline{\Psi}_{sls}^{\quad xy} = \overline{\iota}_{ss}^{xy} L_{sl} \tag{3}$$

$$\overline{\Psi}'_{rlS}{}^{xy} = \overline{\iota}'^{xy}_{rS}L_{rl} \tag{4}$$

The current sensors in the rotor circuit measure the rotor current in the rotor oriented frame (i_{rR}^{xy}) , the measured current should be transformed into the stationary frame $(\vec{\iota}_{rS}^{xy})$, so this should be also considered during the transformation apart from the turn ratio (a_N) of the machine. Transformation is based on the rotor angle and written in (5):

$$\vec{t}_{rS}^{xy} = a_N i_{rR}^{xy} e^{j\alpha_{rmech}} \tag{5}$$

The stator resistance, stator leakage inductance and rotor leakage inductance are considered to be known.



Fig. 2 Closed-loop induction machine motor model [13], where "r" index refers to rotor, "mr" index refers to magnetic field

B. Traditional Rotor Flux Estimation

Equation (6) is the rotor voltage equation in the rotor flux oriented reference frame:

$$0 = R_r \bar{\iota}_{r\Psi r} + \frac{d|\bar{\Psi}_{r\Psi r}|}{dt} + j(\omega_{mr} - \omega_r)|\bar{\Psi}_{r\Psi r}|$$
⁽⁶⁾

Knowing that the rotor flux space vector in rotor reference frame is:

$$|\overline{\Psi}_{r\Psi r}| = L_m |\overline{\iota}_{m\Psi r}| \tag{7}$$

$$\bar{\iota}_{r\Psi r} = \frac{|\iota_{m\Psi r}| - \iota_{s\Psi r}}{1 + \frac{L_{rl}}{L_m}} \tag{8}$$

If we substitute (7), (8) into (6) and use $T_r = \frac{L_r}{R_r}$

$$T_r \frac{d|\bar{\iota}_{m\Psi r}|}{dt} + |\bar{\iota}_{m\Psi r}| = \bar{\iota}_{s\Psi r} - j(\omega_{mr} - \omega_r)T_r|\bar{\iota}_{m\Psi r}|$$
⁽⁹⁾

As we solve the (9) we get the result for the components:

$$T_r \frac{d|\bar{\iota}_{m\Psi r}|}{dt} + |\bar{\iota}_{m\Psi r}| = \bar{\iota}_{sd}$$
(10)

$$\omega_{slip} = \omega_{mr} - \omega_r = + \frac{i_{sq}}{T_r |\bar{\iota}_m \psi_r|}$$
(11)

C. Rotor Resistance Tuning

Considering two different given cases either $\Psi_{rSRef}^{'xy}$ is leading to $\Psi_{rS}^{'xy}$ or $\Psi_{rSRef}^{'xy}$ is lagging to $\Psi_{rS}^{'xy}$ it is possible to create a controller which is suitable to adjust the resistance used in the machine model. The difference of two calculated flux vectors is caused by the deviation between the real and estimated rotor resistance.

- if the estimated rotor resistance is higher than the real value the angle between the reference and estimated rotor flux is negative. It is a consequence of the fact that it can be considered as if the slip was reduced so the angle between the stator and rotor flux would be increased.
- the case when the angle between the reference and estimated rotor flux is positive can be considered as a complement of the previous case.



Fig. 3 Estimated rotor resistance

A PI controller is used in the algorithm to get the correct estimated value of the rotor resistance (R_{rest}) which is written into the machine model according to following.

- PI controller should increase the Rrest if the $\alpha_{Ref} \alpha_r^*$ angle error is positive.
- PI controller should reduce the Rrest if the $\alpha_{Ref} \alpha_r^*$ angle error is negative.



Fig 4. Park vector path of Rotor and Stator flux



Fig. 5 Park vector of Rotor and Stator flux at two different time.

SIMULATION AND SIGNAL POST-PROCESSING

The simulation aimed to emulate a motor test bench on which the measurement was done. The power circuit model is based on the SimPower System Toolbox and signal pre-processing is done by standard Simulink blocks.

The SRIM test bench with 55kW shaft power was chosen as a subject of the simulation. Previously, the motor parameters were identified by the traditional blocked–rotor and open–circuit measurement. The SimPower system model was parametrized by these parameters except for the iron loss resistance which was neglected.

$$R_S = 32 \ m\Omega R_R = 95.5 \ m\Omega L_{sl} = 340 \ uHL_{rl} = 340 \ uH$$

 $L_m = 12.3 \ mH$

To verify the robustness of the algorithm, the initial value of the rotor resistance is set to 50% of the real value. Despite a huge error in the initial value of the rotor resistance (50%) the algorithm finds the right resistance value of the tested machine.

In Fig. 3 the transient of the rotor resistance can be seen. It resulted in the motor model transient because of the zero initial conditions and the transient of the PI controller in the rotor resistance estimator loop. The track of the rotor flux Park vector can be seen in Fig. 4 and Fig. 5. It can be seen that the machine model is demagnetized at the beginning so the magnitude of the flux is initially zero.

CONCLUSION

The article introduced the MotIL approach to help motor control developers with verification. MotIL, HIL and PHIL solutions can be integrated into one developing workflow in which the MotIL is the physical platform of the system verification including all relevant physical effects. The MotIL can help the real time model development for the HIL and PHILs. A rotor resistance monitoring algorithm was introduced as a MotIL case study. Thanks to the recommended monitoring method, the test bench operation can be monitored externally without injection of the sensing signals. Finally, the performance of the recommended solution is verified by the Matlab based simulation.

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